• CLASS TEST •						SI.:01SK_EC_ABCDEFG_10624				
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ELECTRONIC DEVICES										
ELECTRONICS ENGINEERING										
	Date of Test : 10/06/2024									
AN	SWER KEY	>								
1.	(a)	7.	(c)	13.	(a)	19.	(b)	25.	(c)	
2.	(c)	8.	(c)	14.	(a)	20.	(d)	26.	(b)	
3.	(b)	9.	(a)	15.	(c)	21.	(a)	27.	(b)	
4.	(c)	10.	(b)	16.	(b)	22.	(b)	28.	(a)	
5.	(a)	11.	(b)	17.	(b)	23.	(c)	29.	(a)	
6.	(a)	12.	(b)	18.	(b)	24.	(c)	30.	(c)	

# **DETAILED EXPLANATIONS**

#### 1. (a)

Zener breakdown voltage is less because in higher doping region, depletion layer width is small and a small reverse voltage is able to break the covalent bond and gives sudden increase in current.

Hence, zener breakdown voltage  $V_1$  corresponds to point A.

#### 2. (c)

We know that,

Collector current,  $I_{C} = \beta I_{B} + (1 + \beta) I_{CO}$  $I_{C} = \beta I_{B} + \beta I_{CO} + I_{CO}$  $\beta = \frac{I_C - I_{CO}}{I_B + I_{CO}}$ 

....

 $\alpha = \frac{\beta}{1+\beta} = \frac{\frac{I_C - I_{CO}}{I_B + I_{CO}}}{1 + \frac{I_C - I_{CO}}{I_B + I_{CO}}} = \frac{I_C - I_{CO}}{I_B + I_{CO} + I_C - I_{CO}}$ 

...

but,

 $\alpha = \frac{I_C - I_{CO}}{I_C + I_P}$ 

### 3. (b)

Given, Oxide thickness  $(t_{ox}) = 50 \text{ nm}$  $C = \frac{\epsilon_{ox}}{t_{ox}} = \frac{3.45 \times 10^{-11}}{50 \times 10^{-9}} \,\mathrm{F/m^2} = 6.9 \times 10^{-4} \,\mathrm{F/m^2}$ Capacitance per unit area,  $C = 69 \, \mathrm{nF}/\mathrm{cm}^2$ ...

4. (c)

> Given, Collector current,  $I_C = 1 \text{ mA}$ Since,  $I_C = 1 \text{ mA} > 0 \Rightarrow BJT$  not in cut-off region. Emitter current  $I_F = 1.2 \text{ mA}$  $I_E = I_B + I_C$ But  $I_{B} = I_{E} - I_{C} = 0.2 \text{ mA}$  $\Rightarrow$  $\beta = \frac{I_C}{I_P} = \frac{1 \text{ mA}}{0.2 \text{ mA}} = 5 < \beta_{\min}$

: BJT is in saturation region.

### 5. (a)

The conductors have the positive temperature coefficient of resistance. The conductors have almost linear increase in resistivity.

### 6. (a)

In a forward biased pn-junction diode, the current flow is due to diffusion of majority carriers and recombination of minority carriers.

7. (c)  

$$\frac{D_n}{\mu_n} = \frac{D_p}{\mu_p} = V_T \quad \text{(Eienstein's equation)}$$

$$\Rightarrow \qquad \frac{48}{\mu_n} = \frac{12}{\mu_p}$$

$$\Rightarrow \qquad 4 = \frac{\mu_n}{\mu_p}$$

$$\mu_p + \mu_n = 100 \qquad \text{(given)}$$

$$\frac{5\mu_p = 100}{\mu_n = 80 \text{ cm}^2/\text{V-sec}}$$
8. (c)  
Given, resistivity,  $\rho = 1.5 \ \Omega\text{-cm}$   
Hall coefficient,  $R_H = -1250 \ \text{cm}^3/\text{C}$   
Since,  $R_H$  is negative, the charge carriers are electrons.  
Mobility,  $\mu_e = \sigma |R_H|$   

$$= \frac{1}{p} |R_H| = \frac{1}{1.5} \times 1250$$

$$\therefore \qquad = 833 \ \text{cm}^2/\text{V-sec}$$
9. (a)  
 $\alpha = \beta^* \gamma$   
 $\gamma = \text{emitter injection efficiency,  $\gamma = \frac{98}{100} = 0.98$   
 $\beta^* = \text{base transport factor, } \beta^* = \frac{99 - 1.98}{99} = 1 - \frac{2}{100} = 0.98 \times 0.98 = 0.9604$ 
10. (b)  
 $\lambda \le \frac{1.24}{E_g(\text{in eV})} \mu m$   
 $\lambda_{(\text{max})} = \frac{1.24}{2.5} \mu m = 0.496 \ \mu m = 4960 \ \text{\AA}$$ 

11. (b)

Given, MOSFET is operated in saturation region and channel length modulation is present,

$$\therefore \text{ Drain current,} \qquad I_D = \mu_n C_{ox} \frac{W}{2L} (V_{GS} - V_T)^2 [1 + \lambda V_{DS}] \qquad \dots (i)$$

Drain to source conductance,

$$g_{ds} = \frac{\partial I_D}{\partial V_{DS}} = \mu_n C_{ox} \frac{W}{2L} (V_{GS} - V_T)^2 (\lambda) \qquad \dots (ii)$$

...(i)

From equation (i), we can write,

$$\frac{I_D}{1+\lambda V_{DS}} = \mu_n C_{ox} \frac{W}{2L} (V_{GS} - V_T)^2$$

We can re-write equation (ii) as,

 $g_{ds} = \frac{I_D}{1 + \lambda V_{DS}} \cdot \lambda$ 

 $g_{ds} = \frac{I_D}{\frac{1}{\lambda} + V_{DS}}$ 

or,

## 12. (b)

The current,

$$I_{D} = \frac{3.4 \text{ V}}{R}$$
  
But it is given, to produce 1 mcd, $I_{D(\min)} = 4 \text{ mA}$   
So,  
$$\frac{3.4 \text{ V}}{R} \ge 4 \text{ mA}$$
$$R \le \frac{3.4}{4} \text{ k}\Omega = 850 \Omega$$
$$R_{(\max)} = 850 \Omega$$
(a)  
$$P_{n} = P_{n0} e^{0.5/V_{f}}$$

$$P_{n0} = \frac{n_i^2}{n_{n0}} = \frac{10^{20}}{5 \times 10^{15}} = 2 \times 10^4 \,\mathrm{cm}^{-3}$$
$$P_n = 2 \times 10^4 \,e^{20} \approx 10 \times 10^{12} \,\mathrm{cm}^{-3}$$

 $I_D = \frac{5 \,\mathrm{V} - 1.6 \,\mathrm{V}}{R}$ 

## 14. (a)

13.

The Fermi potential, 
$$\phi_{fp} = \frac{kT}{q} \ln \left[ \frac{N_a}{n_i} \right]$$

$$\phi_{fp} = 0.026 \ln \left[ \frac{10^{16}}{1.5 \times 10^{10}} \right]$$
  
 $\phi_{fp} = 0.35 \text{ V}$ 

The maximum space charge width

$$x_{d \text{(max)}} = \left[\frac{4\varepsilon_{si}\phi_{fp}}{eN_a}\right]^{1/2} = \left[\frac{4 \times 1.04 \times 10^{-12} \times 0.35}{1.6 \times 10^{-19} \times 10^{16}}\right]^{1/2} \text{cm}$$
$$= 3.02 \times 10^{-5} \text{ cm}$$
$$x_{d \text{(max)}} = 0.302 \,\mu\text{m}$$

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# 15. (c)

Given current equation of a PN junction diode is,

$$\begin{split} I(T) &= I_s(T_0) \exp \left[ 0.057 \left( T - T_o \right) \right] ; T_0 = 300 \text{ K} \\ I(T) &= I_s (300) \exp \left[ 0.057 \left( T - 300 \right) \right] \qquad \dots (i) \end{split}$$

### At T = 300 K,

$$I(300) = I_s (300) \exp [0.057 (300 - 300)]$$
  

$$I(300) = I_s (300) \qquad ...(ii)$$

It is given that, at a temperature (assume  $T_1$ ) the current is twice its value at T = 300 K. The new current at  $T_1$  is

$$I(T_1) = 2I$$
 (300) ...(iii)

$$I(T_1) = I_s (300) \exp [0.057 (T_1 - 300)]$$
 ...(iv)

Substitute equation (iii) in equation (iv),

$$2I(300) = I_s(300) \exp [0.057 (T_1 - 300)]$$

From equation (ii),

$$I (300) = I_s (300)$$
  

$$2I_s (300) = I_s (300) \exp [0.057 (T_1 - 300)]$$
  

$$2 = \exp [0.057 (T_1 - 300)]$$
  

$$In 2 = 0.057 (T_1 - 300)$$
  

$$T_1 = 312.16 \text{ K}$$
  

$$T_1 = 313 \text{ K}$$

:.

But,

### 16. (b)

Given that,

Reverse saturation current density, $J_s = 3.6 \times 10^{-11} \text{ A/cm}^2$ Photo current density, $J_L = 15 \text{ mA/cm}^2$ The open-circuit voltage of a solar cell is,

$$V_{\rm OC} = V_t \ln \left[ 1 + \frac{J_L}{J_s} \right] = 0.026 \ln \left[ 1 + \frac{15 \times 10^{-3}}{3.6 \times 10^{-11}} \right]$$
$$V_{\rm OC} = 0.516 \text{ V}$$

## 17. (b)

Diffusion potential (or) built in potential,

 $V_{bi}$  = Area under the electric field distribution curve = Area under given curve (which resembles triangle)

$$V_{bi} = \frac{1}{2} \times \text{Base} \times \text{height} = \frac{1}{2} \times (W_P + W_N) \times (-E)$$

magnitude of diffusion potential

$$|V_{bi}| = \left|\frac{1}{2} \times (W_P + W_N) \times (-E)\right|$$
$$= \frac{1}{2} \times 4 \,\mu\text{m} \times 15 \times 10^4 \text{ V/m}$$
$$|V_{bi}| = 0.3 \text{ V}$$

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# 18. (b)

As drain is connected to gate for both MOSFETs, they will be in saturation mode of operation. i.e.,  $I_{D1} = I_{D2}$ 

$$K_{1}(V_{GS1} - V_{T})^{2} = K_{2}(V_{GS2} - V_{T})^{2}$$
  
Since,  $K \propto W$ ;  $K_{2} = 2K_{1}$   
 $K_{1}(V_{GS1} - V_{T})^{2} = 2K_{1}(V_{GS2} - V_{T})^{2}$   
But  
 $V_{GS1} = 5 - V_{0}$   
 $V_{GS2} = V_{0}$   
 $(5 - V_{0} - 1.5)^{2} = 2(V_{0} - 1.5)^{2}$   
 $3.5 - V_{0} = \sqrt{2}(V_{0} - 1.5)$   
 $3.5 - V_{0} = \sqrt{2}(V_{0} - 2.12)$   
 $2.414V_{0} = 5.62$   
 $V_{0} = 2.33$  V

19. (b)

$$\begin{split} C_T &\propto \ \frac{1}{\sqrt{V_{bi} + V_{RB}}} \\ \frac{C_{T1}}{C_{T2}} &= \ \frac{\sqrt{V_{bi} + V_{RB2}}}{\sqrt{V_{bi} + V_{RB1}}} \\ V_{RB1} &= \ 0 \ \text{V}, \qquad C_{T1} &= \ 1 \ \mu\text{F}. \\ V_{RB2} &= \ -6 \ \text{V}, \qquad C_{T2} &= \ 0.5 \ \mu\text{F}. \end{split}$$

at at

$$V_{RB2} = -6 \text{ V},$$

$$\frac{C_{T1}}{C_{T2}} = \frac{\sqrt{V_{bi} + V_{RB2}}}{\sqrt{V_{bi} + V_{RB1}}}$$

$$\frac{1}{0.5} = \sqrt{\frac{-6 + V_{bi}}{V_{bi}}}$$

$$V_{bi} = -2 \text{ V}$$

20. (d)

*.*..

We know that,

Diode voltage  $V_D$  decreases by 2.5 mV per 1°C rise in temperature. Given, temperature,  $T_1 = 20$ °C

$$\frac{\Delta V_D}{\Delta T} = -2.5 \text{ mV/}^{\circ}\text{C}$$
$$\frac{V_{D2} - V_{D1}}{(T_2 - 20)} = -2.5 \times 10^{-3}$$
$$\therefore \qquad \frac{(600 - 700) \times 10^{-3}}{-2.5 \times 10^{-3}} = T_2 - 20$$
$$\therefore \qquad T_2 = 60^{\circ}\text{C}$$

# 21. (a)

We know that,

Potential function, 
$$\phi_s = V_T \ln\left(\frac{n_0}{n_i}\right)$$

or,

 $\phi_s = V_T \ln\left[\frac{N_D}{n_i}\right]$ 

Let potential at 1  $\mu$ m distance is  $\phi_{s1'}$ 

$$\therefore \qquad \qquad \phi_{s1} = V_T \ln\left[\frac{N_{D2}}{n_i}\right]$$

$$\phi_{s1} = V_T \ln\left[\frac{10^{16}}{n_i}\right]$$

Let potential at 2  $\mu$ m distance is  $\phi_{s2'}$ 

$$\therefore \qquad \qquad \phi_{s2} = V_T \ln\left[\frac{N_{D1}}{n_i}\right] \qquad \qquad \dots (ii)$$

It is given magnitude of potential difference, i.e.,  $|\phi_{s1} - \phi_{s2}| = 0.12 \text{ V}$ 

$$V_T \ln \left[ \frac{N_{D2}}{N_{D1}} \right] = 0.12 \text{ V}$$
  

$$\therefore \qquad \frac{N_{D2}}{N_{D1}} = e^{\frac{0.12}{0.026}}$$
  

$$\Rightarrow \qquad N_{D1} = 9.89 \times 10^{13} \text{ cm}^{-3}$$

22. (b)

Since two MOSFETs are operated in saturation mode,

$$I_{DN} = \mu_n C_{ox} \frac{W_n}{2L} (V_{GSN} - V_{TN})^2$$
$$I_{DN} = \mu_n C_{ox} \frac{W_n}{2L} (V_{OV})^2$$

or,

where, overdrive voltage  $V_{OV} = V_{GSN} - V_{TN}$ 

Similarly,  $I_{DP} = \mu_p C_{ox} \frac{W_p}{2L} (V_{GSP} - V_{TP})^2$ 

$$I_{DP} = \mu_p C_{ox} \frac{1}{2L} (V_{GSP} - V_{TP})$$
$$I_{DP} = \mu_p C_{ox} \frac{W_p}{2L} (V_{OV})^2$$

or,

where, overdrive voltage  $V_{OV} = V_{GSP} - V_{TP}$ given,  $I_{DN} = I_{DP}$ 

$$\mu_n C_{ox} \frac{W_n}{2L} (V_{OV})^2 = \mu_p C_{ox} \frac{W_p}{2L} (V_{OV})^2$$

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...(i)

$$\therefore \qquad \frac{W_n}{W_p} = \frac{\mu_p}{\mu_n} = 0.4$$

In the given question, it is asking that,

$$\frac{W_p}{W_n} = \frac{1}{0.4} = 2.5$$

23. (c)

24. (c)

Fill factor of solar cell is,

F.F. = 
$$\frac{\text{Maximum power obtained}}{V_{oc} \times I_{sc}}$$
$$0.65 = \frac{65 \times 10^{-3}}{V_{oc} \times I_{sc}}$$
$$V_{oc} \times I_{sc} = \frac{65 \times 10^{-3}}{0.65} = 100 \text{ mW}$$

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: Option (c) satisfies the result ( $V_{oc} \times I_{sc} = 40 \text{ mA} \times 2.5 \text{ V} = 100 \text{ mW}$ )

## 25. (c)

Given,  $\mu_n C_{ox} \frac{W}{L} = 1.5 \times 10^{-3} \text{ A/V}^2$   $V_T = 0.65 \text{ V}$   $V_{GS} = 4 \text{ V}$   $V_{DS} = 6 \text{ V}$ Power dissipation in the MOSFET is, P = V = V

$$P = V_{DS} \times I_{DS}$$

where,  $I_{DS}$  is drain to source saturation current. Since the MOSFET is operating in saturation region,

$$I_{DS} = \frac{1}{2} \mu_n C_{ox} \frac{W}{L} (V_{GS} - V_T)^2$$
$$= \frac{1}{2} \times 1.5 \times 10^{-3} (4 - 0.65)^2$$
$$I_{DS} = 8.42 \times 10^{-3} \text{ A}$$
Power dissipation, 
$$P = 6 \times 8.42 \times 10^{-3}$$
$$\therefore \qquad P = 50.50 \text{ mW}$$

26. (b)

The steady state increase in conductivity,

$$\Delta \sigma = q(\mu_n + \mu_p)(\delta p)$$

In steady state,  $\delta p = g' \tau_{po}$ 

where, g' is the uniform generation rate.

$$\therefore \qquad \Delta \sigma = q(\mu_n + \mu_p) (g' \tau_{po})$$



$$2 = (1.6 \times 10^{-19}) (8500 + 400)g' \times 10^{-7}$$
  
$$\therefore \qquad g' = 1.404 \times 10^{22} \text{ cm}^{-3} \text{ s}^{-1}$$

27. (b)

for

for

 $\Rightarrow$ 

$$J_{p} = -q D_{p} \frac{dp}{dx}$$

$$x \le w,$$

$$\frac{dp}{dx} = \frac{p(0) - p_{0}}{0 - w} = \frac{-(p(0) - p_{0})}{w}$$

$$J_{p} = \frac{q D_{p} (p(0) - p_{0})}{w} = \text{positive constant}$$

$$x \ge w,$$

$$\frac{dp}{dx} = 0$$

$$J_{p} = 0$$

28. (a)

$$q \phi(x) = E_F - E_i(x)$$

$$q \phi(0) = E_F - E_i(0) = kT ln \left(\frac{N_D(0)}{n_i}\right)$$

$$\phi(0) = \frac{kT}{q} ln \left(\frac{N_D(0)}{n_i}\right)$$

$$\phi(x = 5\mu m) = \frac{kT}{q} ln \left(\frac{N_D(5\mu m)}{n_i}\right)$$

$$V_0 = \phi(0) - \phi(x = 5\mu m) = \frac{kT}{q} ln \left(\frac{N_D(0)}{N_D(5\mu m)}\right)$$

$$= 0.026 ln \left(\frac{10^{20}}{10^{15}}\right) V \simeq 0.3 V$$

29. (a)

 $\phi_s$  = surface potential Electric field  $\epsilon_{si}$ ε<sub>ox</sub>  $= \frac{1}{2}E_s W_{dep}$  $E_{ox}$ Es  $E_{s} = \frac{2\phi_{s}}{W_{dep}} = \frac{2 \times 0.035}{0.4} \text{ V/}\mu\text{m}$ SiO<sub>2</sub> Si •  $\phi_s$ = Area  $E_s = 0.175 \text{ V/}\mu\text{m}$ Boundary condition of electrostatic field should be satisfied at the interface W<sub>dep</sub>  $\varepsilon_{ox}E_{ox} = \varepsilon_{si}E_s$ SiO<sub>2</sub> - Si interface

$$E_{ox} = \frac{\varepsilon_{si} E_s}{\varepsilon_{ox}} = 3E_s$$
$$E_{ox} = 0.525 \text{ V/}\mu\text{m}$$

So,

$$E_{ox} = \frac{V_{ox}}{t_{ox}}$$

$$V_{ox} = E_{ox}t_{ox} = 0.525 \times 1.8 \text{ V} = 0.945 \text{ V}$$

$$V_G = V_{ox} + \phi_s = 0.98 \text{ V}$$

30. (c)

$$\vec{F}_m = q\vec{v}_d \times \vec{B} = q\mu_p \vec{E}_{applied} \times \vec{B} = \frac{q\mu_p V_x}{L} \hat{x} \times (-10\hat{z})$$

$$= \frac{100 q\mu_p}{L} \hat{y}$$

$$\vec{F}_{ei} = \text{force due to induced Hall electric field} = -\vec{F}_m$$

$$= \frac{100 q\mu_p}{L} (-\hat{y}) = q\vec{E}_{ind}$$

$$\vec{E}_{ind} = \frac{100 \mu_p}{L} (-\hat{y})$$

As  $\vec{E}_{ind}$  in  $(-\hat{y})$  direction,  $V_{H}$  is +ve

Force due to magnetic field,

$$V_{\rm H} = W \left| \vec{E}_{\rm ind} \right| = \frac{W(100)\mu_p}{L} = \frac{W(100)(500 \times 10^{-4})}{2W} = 2.5 \, {\rm V}$$