

- 1. This question paper contains **30** objective questions. **Q.1-10** carry one mark each and **Q.11-30** carry two marks each.
- 2. Answer all the questions.
- 3. Questions must be answered on Objective Response Sheet (**ORS**) by darkening the appropriate bubble (marked **A**, **B**, **C**, **D**) using HB pencil against the question number. Each question has only one correct answer. In case you wish to change an answer, erase the old answer completely using a good soft eraser.
- 4. There will be **NEGATIVE** marking. For each wrong answer **1/3rd** of the full marks of the question will be deducted. More than one answer marked against a question will be deemed as an incorrect response and will be negatively marked.
- 5. Write your name & Roll No. at the specified locations on the right half of the ORS.
- 6. No charts or tables will be provided in the examination hall.
- 7. Choose the **Closest** numerical answer among the choices given.
- 8. If a candidate gives more than one answer, it will be treated as a **wrong answer** even if one of the given answers happens to be correct and there will be same penalty as above to that questions.
- 9. If a question is left blank, i.e., no answer is given by the candidate, there will be **no penalty** for that question.

DO NOT OPEN THIS TEST BOOKLET UNTIL YOU ARE ASKED TO DO SO

Q.No. 1 to Q.No. 10 carry 1 mark each

Q.1 The average value of the periodic signal shown below is



Q.2 In the circuit shown in figure, the switch *S* is closed at time (t = 0). The voltage across the inductance ($V_t(t)$) at $t = 0^+$, is



Q.3 Consider the circuit shown below



The current flowing through 50 V source is (a) 30 A (b) 20.66 A

- (c) 5.48 A (d) 2.34 A
- **Q.4** Determine the voltage that must be applied at x y terminal such that the voltage across 4 Ω is 5 V.



Q.5 For the parallel RC circuit shown below, if R = 1 Ω , and C = 1 F. The value of $Z_{21}(s)$ will be



- **Q.6** If the current in a 20 Ω resistor is given by $i = 4 + 5 \sin \omega t - 3 \cos 3 \omega t$ The power consumed by the resistor is
 - (a) 1 kW (b) 1.865 W (c) 660 W (d) 720 W
- **Q.7** Consider the circuit shown in the figure below:







If the $I_o(j\omega)$ is current flowing in the circuit for particular value of angular frequency ω , then the

value of	$\frac{I_o(j)}{I_o(5j)}$	is equal to)
(a) 2.1		(b)	3.6
(c) 4.2		(d)	8.8

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Q.9 Consider the circuit shown in the figure below:



(a) -0.25 s (b) -0.5 s (c) -0.75 s (d) -1 s

Q.10 Consider the circuit shown in figure below:



Assume
$$V_c(0^-) = 5$$
 V. If $v_c(t) = \frac{5}{e}V$ at $t = 0.1$

sec, then the value of C is

(a) $1.2 \,\mu\text{F}$ (b) $1.5 \,\mu\text{F}$ (c) $2.5 \,\mu\text{F}$ (d) $5 \,\mu\text{F}$

Q. No. 11 to Q. No. 30 carry 2 marks each



If the switch is at position '1' for a long time and at (t = 0) it is moved to position '2', then the current *i*(*t*) for *t* > 0 will be

- (a) $(5e^{-t} + 25e^{t}) A$ (b) $(5 + 25te^{-t}) A$ (c) $(5te^{-t}) A$ (d) $(25te^{-t} + 5) A$
- Q.12 The time constant of the circuit shown below will be



Q.13 For the circuit shown below, the network N has

Z parameter matrix of $\begin{bmatrix} 4 & 8 \\ 1 & 3 \end{bmatrix}$. If the transformer

used in the circuit is ideal, then the power delivered to the 1 Ω resistance will be



Q.14 The hybrid parameter of the network *N* shown below are $h_{11} = 2 \Omega$, $h_{12} = 4$, $h_{21} = -5$ and $h_{22} = 2 \Im$. The supply voltage, if the power dissipated in the load resistor $R_L(=4 \Omega)$ is 25 W and R_s is given by 2 Ω is



Q.15 For a series RLC circuit the magnitude of frequency at which the drop across the capacitor is maximum will be

(d) 58 V

(a)
$$\sqrt{\frac{1}{LC} - \left(\frac{RC}{L}\right)^2}$$
 rad/sec

(b)
$$\frac{1}{\sqrt{LC}}$$
 rad/sec

(c) 43 V

(c)
$$\sqrt{\frac{1}{LC} - \frac{R^2}{2L^2}}$$
 rad/sec

(d)
$$\sqrt{\frac{1}{LC} - R^2 C^2}$$
 rad/sec

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Q.16 In below figure the capacitor initially has a charge of 10 *C*. The current in the circuit after two seconds the switch *S* is closed will be



Q.17 Consider the circuit shown below



The maximum power absorbed by the load is

- (a) 390.625 W (b) 400.125 W
- (c) 500.125 W (d) 350.625 W
- Q.18 Consider the circuit shown below



The Thevenin's equivalent resistance across the terminals *a* and *b* is

(a)	10 Ω	(b) 20	Ω (
(01)		(/ _ `		

(c) 30 Ω	(d)	60Ω
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Q.19 A series resonance circuit has a source frequency of 5 kHz and source impedance of $(2 + j4) \Omega$. The load impedance being $(10 - j X_C) \Omega$. The value of *C* such that the power consumed by the resistor is maximum is

(a)	1.45 μF	(b)	7.95 μF
(c)	2.84 uF	(d)	9.87 uF

Q.20 The voltage applied to series *R*-*L* circuit at $t = t_0$ sec is given by $10\cos\left(\frac{t}{2} + 45^\circ\right)$ V. The value of resistance is 1 Ω and the value of inductance

is 2 H. The value of t_0 to obtain the transient free response is

(a) $\frac{\pi}{4}$ sec (b) $\frac{\pi}{2}$ sec

(c) $\pi \sec$ (d) $2\pi \sec$

Q.21 Find the transfer function $\frac{V_o(s)}{I(s)}$ for the network shown



(c)
$$\frac{8s(s+1)}{2s^2+6s+1}$$
 (d) $\frac{2s(s+1)}{2s^2+6s+1}$

Q.22 For the two port network *N* shown below in figure *Z* parameter matrix is $\begin{bmatrix} 2 & 1 \\ 1 & 4 \end{bmatrix}$. The value of I_1 in

Amperes is 3Ω 141 V + V_1 V_2 K_2 K_1 V_2 K_1 V_2 V_2 K_1 V_2 K_1 V_2 K_1 V_2 V_2 V_1 V_2 V_2 V_1 V_2 V_2 V_2 V_1 V_2 V_2 $V_$

(c) 24

Q.23 In the figure shown, all elements used are ideal. For time t < 0, S_1 remain closed and S_2 open. At t = 0, S_1 is opened and S_2 is closed. If the voltage V_{C_2} across the capacitor C_2 at $t = 0^-$ is zero, the voltage across the capacitor combination at t > 0 will be

(d) 28



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Q.24 Consider the circuit shown in the figure below:



The value of I_{α} is equal to

- (a) 5.831∠149.03° (b) 2∠180°
- (c) 6.913∠132.1° (d) 2.915∠-30.96°
- Q.25 For the circuit shown below the equivalent impedance seen across the terminals *A* and *B* is



Q.26 In the figure shown below, the switch *S* is moved from position 1 to 2 at time t = 0. Just before the switch is thrown, the initial conditions are $i_L(0^-) = 2$ A and $V_c(0^-) = 2$ V. Then the current i(t) for t > 0 is



(a)	(<i>e</i> ^{<i>t</i>} - <i>e</i> ^{-2t}) A	(b)	$(e^{-t} - e^{-2t}) A$
(C)	$(e^{-t} + e^{-2t}) A$	(d)	$(e^t + e^{2t}) A$

Q.27 The resonant frequency of a series *RLC* circuit is 1.5 MHz with the resonating capacitor set at 150 pF. If the bandwidth is 10 kHz, then the effective resistance of the circuit would be (approximately)

(a)	1.25 Ω	(b)	2.25	Ω
(C)	3.25 Ω	(d)	4.71	Ω

Q.28 A voltage signal v(t) is applied to a capacitor with capacitance equal to 10 μ F. The voltage wave is shown in the figure below. Which of the following plot is correct for the current *i*(*t*) through the capacitor?







Assuming 'a' to be a positive non zero number, then which of the following statement is correct?(a) Z-parameter does not exists for all values of 'a'

- (b) Z-parameter does not exists for 'a' = 3
- (c) Z-parameter does not exists for 'a' = 8
- (d) Z-parameter exists for all positive value of 'a'





(a) 2.63Ω (b) 6.33Ω (c) 21.4Ω (d) 19Ω

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AN	SWER KEY	>							
1.	(a)	7.	(b)	13.	(b)	19.	(b)	25.	(b)
2.	(b)	8.	(a)	14.	(d)	20.	(c)	26.	(c)
3.	(c)	9.	(a)	15.	(c)	21.	(a)	27.	(d)
4.	(b)	10.	(c)	16.	(a)	22.	(c)	28.	(c)
5.	(d)	11.	(c)	17.	(a)	23.	(b)	29.	(d)
6.	(c)	12.	(b)	18.	(d)	24.	(d)	30.	(a)

DETAILED EXPLANATIONS

1. (a)

The average value of periodic signal can be calculated by considering one time period

$$= \frac{\text{Total area under the graph for one period}}{T_0}$$

Total area under the graph for one period = Area 1 + Area 2 + Area 3 + Area 4
here Area 1 = Area 2 = Area 3 = Area 4 = 4
and $T_0 = 8 \sec$
Average value = $\frac{4+4+4+4}{8} = \frac{16}{8} = 2$
 $\int_{0}^{1} (2) (3) = \int_{0}^{1} (2) (3) = \int_{0}^{1} (2) (3) = \int_{0}^{1} (3) \int_{0}^{1} (3) = \int_{0}^{1} (3) \int_{0}$

2. (b)

Before closing the switch, the circuit was not energized, therefore, current through inductor and voltage across capacitor are zero.

After closing the switch, at $t = 0^+$ inductor acts as open-circuit and capacitor acts as short-circuit. Equivalent circuit at $t = 0^+$



3. (c)

using source transformation



applying KCL at V

$$\frac{V-50}{5} + \frac{V-20}{2} + \frac{V-10}{3} = 0$$

$$\frac{6V-300+15V-300+10V-100}{31V-700} = 0$$
or
$$V = \frac{700}{31} = 22.58 V$$

$$i = \frac{50-V}{5}$$

$$= \frac{50-22.58}{5} = 5.48 A$$

4. (b)



...

5. (d)

Converting the network into s-domain

$$V_2(s) = I_1(s) \left(\frac{R \times \frac{1}{Cs}}{R + \frac{1}{Cs}} \right)$$
$$\frac{V_2(s)}{I_1(s)} = \frac{R}{RCs + 1} = \frac{1}{(s)}$$

 $I_1(s)$ $I_2(s)$ $\frac{1}{Cs}$ $V_1(s)$ $V_2(s)$

or

$$Z_{21}(s) = \frac{V_2(s)}{I_1(s)} = \frac{1}{\left(s + \frac{1}{RC}\right)} = \frac{1}{1(s+1)}$$
$$Z_{21}(s) = \frac{1}{(s+1)}$$

6. (c)

$$P = P_0 + P_1 + P_2$$

= $4^2 \times 20 + \left(\frac{5}{\sqrt{2}}\right)^2 \times 20 + \left(\frac{3}{\sqrt{2}}\right)^2 \times 20$ [$\because I_{\text{rms}} = \frac{I_{\text{max}}}{\sqrt{2}}$]
= $(16 + 12.5 + 4.5) \times 20 = 660 \text{ W}$

7. (b)

The resistance $6\Omega \| 3\Omega$ and $12\Omega \| 4\Omega$ also 1Ω is in series with 5Ω , thus, the circuit can be redrawn as



8. (a

$$Z_{L} = j\omega L = j\Omega$$

$$Z_{C} = \frac{1}{j\omega C} = \frac{1}{j(1)(0.05)} = -j20 \Omega$$

$$\therefore \qquad Z_{eq} = j+2 ||(-j20) = 1.98 + j0.802 \Omega$$
and
$$Z_{L}(5j) = 5j\Omega$$

$$Z_{C}(5j) = -j4 \Omega$$

$$\therefore \qquad Z_{eq}(j5) = j5 + 2 ||(-j4) = 1.6 + j4.2 \Omega$$
Now,
$$I(j\omega) \propto \frac{1}{Z(j\omega)}$$

$$\left|\frac{I_o(j)}{I_o(j5)}\right| = \frac{Z(j5)}{Z(j)} = \left|\frac{1.6 + j4.2}{1.98 + j0.802}\right| = 2.104$$

9. (a)

...



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Now,

...

$$V_{1} = 8I_{1} + V_{1}$$
$$V_{1} = -6V_{x} + V_{x} = -5 V_{x}$$
$$\frac{I_{2}}{V_{1}} = -\frac{1.25}{5} = -0.25 \text{ S}$$

10. (c)

Voltage across capacitor

 $v_c(t) = V_{\text{final}} + (V_{\text{initial}} - V_{\text{final}})e^{-t/RC}$ = 0 + (5 - 0)e^{-t/RC} But given, $v_c(t) = \frac{5}{e} = 5e^{-t/RC}$ $\frac{5}{e} = 5e^{-(0.1/40 \text{ k}\Omega \times C)}$ $\frac{0.1}{40 \text{ k}\Omega \times C} = 1$ ∴ $C = 2.5 \,\mu\text{F}$ (or) $2.5 \times 10^{-6} \text{ F}$

11. (c) At (t = 0⁻)



At $(t = 0^+)$



12. (b)



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13. (b)

Network 'N' can be replaced as



1

Applying KVL at loop (1)

$$\begin{array}{rcl} 20 &=& 4I_1+8I_2\\ \text{also} & & V_2 &=& -9I_2=I_1+3I_2\\ \Rightarrow & & I_1 &=& -12I_2\\ 20 &=& 4(-12\,I_2)+8I_2\\ 20 &=& -40I_2\\ \Rightarrow & & I_2 &=& -0.5 \text{ A}\\ & & I_1 &=& -12(-0.5)=6 \text{ A} \end{array}$$

Now,

$$\frac{I_{\text{primary}}}{I_{\text{secondary}}} = \frac{1}{3}$$

$$\Rightarrow \qquad 3I_{\text{primary}} = I_{\text{secondary}}$$

$$3(-I_2) = I_3$$

$$\Rightarrow \qquad I_3 = 1.5 \text{ A}$$
Power delivered to 1 $\Omega = (I_3)^2 \times R_L = (1.5)^2 \times$

$$= 2.25 \text{ Watts}$$

The *h*-parameter

$$\begin{array}{rcl} V_1 &=& 2I_1 + 4V_2 \\ I_2 &=& -5I_1 + 2V_2 \end{array}$$

Power dissipated in R_{l}

$$\begin{array}{rcl} P_L &=& \displaystyle \frac{V_2^2}{R_L} = 25 \ \mathrm{W} \\ \therefore & V_2 &=& \displaystyle \sqrt{25 \times 4} = 10 \ \mathrm{V} \\ \vdots & & I_2 &=& \displaystyle -I_2 R_L \\ \therefore & & I_2 &=& \displaystyle -2.5 \ \mathrm{A} \\ \text{substituting the values} \\ \text{we get} & V_1 &=& \displaystyle 2I_1 + 40 \\ & & \displaystyle -2.5 &=& \displaystyle -5I_1 + 20 \\ \therefore & & I_1 &=& \displaystyle 4.5 \ \mathrm{A} & \text{and} \end{array}$$



 $V_1 =$

<u>.</u> 000

≠ c

R M

49 V

$$\therefore \qquad I_1 = \frac{V_s - V_1}{2} = 4.5$$
$$\Rightarrow \qquad V_s = 58 \text{ V}$$

15. (c)

For series RLC circuit,

$$Z = \sqrt{R^2 + (X_L - X_C)^2}$$

The circuit current

$$I = \frac{V}{Z}$$

 \therefore The drop across the capacitor C

$$V_C = IX_C = \frac{V}{Z}X_C$$
$$V_C^2 = \frac{V^2 X_C^2}{Z^2} = \frac{V^2}{(\omega C)^2 \left[R^2 + \left(\omega L - \frac{1}{\omega C}\right)^2\right]}$$

i.e.

For maximum voltage drop

$$V_{C \max} = \frac{dV_C}{d\omega} = 0$$
$$= \sqrt{\frac{1}{LC} - \frac{R^2}{2L^2}} \text{ rad/sec}$$

16. (a)

Using KCL

$$100 = R \frac{dQ}{dt} + \frac{Q}{C}$$

$$100 C = RC \frac{dQ}{dt} + Q$$

$$\int_{Q_0}^{Q} \frac{dQ}{100C - Q} = \frac{1}{RC} \int_{0}^{t} dt$$

$$\left(-ln(100C - Q)\right)_{Q_0}^{Q} = \frac{t}{RC}$$

$$ln \frac{(100C - Q)}{(100C - Q_0)} = -\frac{t}{RC}$$

$$100 C - Q = (100 C - Q_0)e^{-t/RC}$$

$$i = \frac{dQ}{dt} = \left(\frac{100C - Q_0}{RC}\right)e^{-t/RC} = \left(\frac{50 - 10}{1}\right)e^{-t/RC}$$

$$i = 40 e^{-t/RC}$$

$$i = 40 e^{-t/RC}$$

$$i = 40 e^{-2} = 5.413 \text{ A}$$

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17. (a) Z_{Th} : j10 Ω 0 Z_{Th} 📥 –j10 Ω 0 10 Ω $Z_{\text{Th}} = \frac{(10+j10)(-j10)}{10+j10-j10} = 10-j10$ *V*_{Th} : $+V_{Th}$ j10 Ω **0**+ V_{Th} ± _j10 Ω 🕻 †) 10∠90° 10 Ω 0-25∠0° $\frac{V_{Th} - 25}{10 + i10} + \frac{V_{Th}}{-i10} = 10 \angle 90^{\circ}$ $\frac{V_{Th}}{10+j10} - \frac{25}{10+j10} - \frac{V_{Th}}{j10} = 10j$ $(V_{\text{Th}})j10 - 250j - 10V_{\text{Th}} - 10jV_{\text{Th}} = -100(10 + 10j)$ $j10 V_{\text{Th}} - 250j - 10 V_{\text{Th}} - 10j V_{\text{Th}} = -1000(1 + j)$ $-10 V_{\rm Th} = -1000 - 1000 \, j + 250 \, j$ $-10V_{\text{Th}} = -1000 - 750 j$ $V_{Th} = 100 + j75$ $Z_L = Z_{Th}^* = 10 + j10$ $P_{\text{max}} = \frac{|V_{Th}|^2}{4\text{Re}\{Z_{Th}\}} = \frac{\left(\sqrt{100^2 + 75^2}\right)^2}{4 \times 10} = 390.625 \text{ W}$ 18. (d)

100 Ω ₩₩ 0.01 V_r 100 Ω 2400 Ω ~~~ Ьb 800 Ω $V = 100 (1 - 2i_x) + 400 (1 - 2i_x - 0.01 V_x) + 800 i_x$ $i_x = 1 \text{ A and } V_x = V$ 5V = 1300 - 1000 = 300V = 60 V*:*.. $R_{\rm TH}$ = 60 Ω and

19. (b)

Given,

$$f_{0} = 5 \text{ kHz}$$

$$Z = Z_{\text{source}} + Z_{\text{load}}$$

$$= 2 + j4 + 10 - jX_{C}$$

$$= [12 + j(4 - X_{C})] \Omega$$
at resonance, imaginary part of Z is zero.

$$\therefore \qquad X_{L} = X_{C}$$

$$\therefore \qquad X_{C} = \frac{1}{\omega_{0}C} = 4$$
or

$$C = 7.95 \,\mu\text{F}$$
As at resonance current is maximum and thus, maximum power transferred to R.

20. (c)

Condition for Transient free response is given by $\omega t_0 = \tan^{-1} \left(\frac{\omega L}{R} \right) - \theta + \frac{\pi}{2}$

here,

$$\omega = \frac{1}{2}, \qquad \theta = 45^{\circ}$$

$$\frac{t_0}{2} = \tan^{-1} \left(\frac{\frac{1}{2} \times 2}{1} \right) - \frac{\pi}{4} + \frac{\pi}{2}$$

$$\frac{t_0}{2} = \frac{\pi}{2}$$

$$t_0 = \pi = 3.1416 \text{ sec}$$

21. (a)

$$I(s) = \frac{I(s)(1+s)}{\frac{1}{s}+2+1+s} = \frac{I(s)(s+1)s}{s^2+3s+1}$$

$$V_o(s) = 2I_1(s) = \frac{I(s)2s(s+1)}{s^2+3s+1}$$

$$\frac{V_o(s)}{I(s)} = \frac{2s(s+1)}{s^2+3s+1}$$

...(*i*)

22. (c)



Applying KVL Loop I

$$\begin{array}{rcl} 3(I_1+I_3)+V_1 &=& 141\\ 3I_1+3I_3+2I_1+I_2 &=& 141\\ 5I_1+I_2+3I_3 &=& 141 \end{array}$$

Loop II

$$V_{2} = (I_{3} - I_{2}) \cdot 6$$

$$V_{2} = I_{1} + 4I_{2}$$

$$I_{1} + 4I_{2} = 6I_{3} - 6I_{2}$$

$$I_{1} + 10I_{2} - 6I_{3} = 0$$
...(*ii*)

Loop III

$$\begin{array}{rcl} 3(I_1+I_3)+3I_3+6(I_3-I_2) &=& 141 \\ && 3I_1-6I_2+12I_3 &=& 141 \\ && & \\ \text{By equation (i), (ii) and (iii)} \\ I_1 &=& 24 \text{ A} & I_2 &=& 1.5 \text{ A} \\ && & I_2 &=& 1.5 \text{ A} \\ \end{array}$$

23. (b)

At $t = 0^- S_1$ is closed. S_2 is open



 C_1 gets charged upto 3 V charge stored in C_1

$$Q_0 = C_1 V = 1 \times 3 = 3C$$

Voltage across C_2 is zero at $t = 0^-$, so no charge stored in C_2 . At t > 0, S_1 is open and S_2 is closed.

charge stored (Q_0) initially in C_1 gets redistributed between C_1 and C_2



Let Charge stored in $C_1 = Q_1$ Charge stored in $C_2 = Q_2$

...(1)

...(2)

According to conservation of charge

 $Q_1 + Q_2 = Q_0 = 3$ Voltage across C_1 = Voltage across C_2

$$\frac{Q_1}{C_1} = \frac{Q_2}{C_2} \implies \frac{Q_1}{1} = \frac{Q_2}{2}$$
$$Q_2 = 2Q_1$$

from (1) and (2)

and

:. Voltage across the combination =
$$\frac{Q_1}{C_1} = \frac{1}{1} = 1$$
 V

 $\begin{array}{rcl} Q_1 &=& 1 \ C \\ Q_2 &=& 2 \ C \end{array}$

24. (d)

Applying KCL on supernode



25. (b)

$$Z_{AB} = \left(\frac{23}{6}\right) + \left[(3+j4)\|(3-j4)\right]$$
$$= \frac{23}{6} + \frac{(3+j4)(3-j4)}{6} = \frac{23+25}{6} = \frac{48}{6} \Omega = 8 \Omega$$
$$Z_{AB} = 8 \Omega$$

26. (c)

...

At t = 0, switch is closed

For t > 0, the circuit in *s*-domain becomes,



Applying KVL, we get,

$$\frac{5}{s} - \frac{2}{s} + 2 = \left(3 + s + \frac{2}{s}\right)I(s)$$

$$I(s) = \frac{2s + 3}{(s+1)(s+2)}$$
Using partial fractions, $I(s) = \frac{1}{(s+1)} + \frac{1}{(s+2)}$
or
$$i(t) = L^{-1}[I(s)] = (e^{-t} + e^{-2t}) A ; \text{ for } t > 0$$

27. (d)

Given,

$$f = 1.5 \text{ MHz}$$

 $C = 150 \text{ pF}$
 $BW = 10 \text{ kHz}$
For series *RLC* circuit,
 $Q = \frac{f_o}{BW} = \frac{1.5 \times 10^6}{10 \times 10^3} = 150$
 $Q = \frac{1}{\omega RC}$
 $\frac{1}{150} = 2\pi \times 1.5 \times 10^6 \times 150 \times 10^{-12} \times R$
 $R = \frac{10^6}{2\pi \times 1.5 \times 150 \times 150} = 4.71 \Omega$

28. (c)

For a capacitor
$$i(t) = \frac{cdv(t)}{dt} = 10 \times 10^{-6} \frac{dv(t)}{dt} \times 10^{3} \text{ A}$$
$$= 10^{-2} \frac{dv(t)}{dt} \text{ A} = 10 \frac{dv(t)}{dt} \text{ mA}$$

29. (d)





For I_2 , we can write

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$$= -\frac{1}{a}V_1 + \frac{2}{a^2}V_2$$

$$\therefore \qquad \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} 2 & -\frac{1}{a} \\ -\frac{1}{a} & \frac{2}{a^2} \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix}$$

For the *Z*-parameter to not exist. |Y| = 0

$$|Y| = 0$$

$$|Y| = \frac{4}{a^2} - \frac{1}{a^2} = \frac{3}{a^2}$$

$$|Y| \neq 0$$

Thus, no such value exist for which |Y| = 0.

30. (a)

From phasor, we can write

$$\tan 30^\circ = \frac{X_C}{R_2}$$

 \Rightarrow

$$R_{2} = X_{C}\sqrt{3} = \frac{\sqrt{3}}{\omega C}$$
$$45^{\circ} = \frac{X_{L}}{R_{1}}$$
$$R_{1} = X_{L} = \omega L$$

 \Rightarrow

$$\tan 45^{\circ} = \frac{L}{R_1}$$

$$R_1 = X_L = \omega L$$

$$R_1 R_2 = \frac{\sqrt{3}}{\omega C} \times \omega L = \frac{L}{C} \sqrt{3}$$

$$R_1 R_2 = \sqrt{3} = 1.732$$

we know

$$\frac{R_1 + R_2}{2} \geq \sqrt{R_1 R_2}$$

as arithmetic mean \geq geometric mean; (for non-negative real numbers)

$$R_1 + R_2 \ge 2\sqrt{\sqrt{3}}$$

 $R_1 + R_2 \ge 2(3)^{1/4}$

Minimum value of $R_1 + R_2 = 2.63 \ \Omega$

