

# **GATE**2023

INSTRUMENTATION ENGINEERING

# Memory based Questions & Solutions

0

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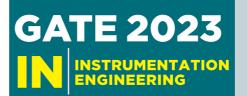
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Exam held on 11<sup>th</sup> Feb, 2023 Forenoon Session





Exam held on:

11-02-2023

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# SECTION - A

# **GENERAL APTITUDE**

Q.1	How many 5-digit number can be formed by using 3, 4, 5 and 6 if exactly One digit
	appearing twice.

Ans. (240)

We have 5 digits out of which 2 are identical.

Number of ways = 
$$\frac{5!}{2!} = \frac{120}{2} = 60$$

Out of four digit exactly one digit appearing twice.

Number of ways =  $4_{C_1} = 4$ 

Total 5-digit number =  $4 \times 60 = 240$ 

**End of Solution** 

- Q.2 Disagree : Protest :: Agree : \_\_\_\_\_
  - (a) Refuse

(b) Recommend

(c) Refute

(d) Pretext

Ans. (b)

**End of Solution** 

- Q.3 The village was nested in a green spot \_\_\_\_ the ocean and the field
  - (a) in

(b) at

(c) through

(d) between

Ans. (d)

End of Solution

- Q.4 Residency is a famous housing complex with many well-established individuals among its residents. A recent survey conducted among the residents of the complex revealed that all of those residents who are well established in their respective fields happen to the academicians. The survey also revealed that most of these academicians are authors of some best-selling books. Based only on the information provided above, which one of the following statements can be logically inferred with certainty?
  - (a) Some authors of best-selling books are residents of the complex who are wellestablished in their fields
  - (b) All academicians residing in the complex are well establishing in their fields.
  - (c) Some residents of the complex who are well established in their fields are also authors of some best-selling books.
  - (d) Some academicians residing in the complex are well established in their fields.

Ans. (d)



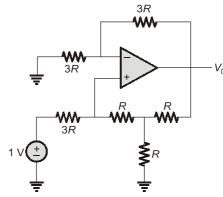
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# **SECTION - B**

# **TECHNICAL**

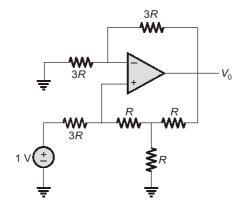
Q.5 For Ideal op-amp shown below:



The value of  $V_0$  is \_\_\_\_\_ V.

Ans. (2)

Assuming op-amp is not saturated.



$$\frac{V_A - 1}{4R} + \frac{V_A}{R} + \frac{V_A - V_O}{R} = 0$$

$$\frac{V_A - 1}{4} + V_A + V_A - V_O = 0$$

$$9V_A - 1 - 4V_o = 0$$

$$V_A = \frac{1}{9} + \frac{4}{9}V_O$$

 $\therefore$  Current flowing through  $3R \Omega$ .

$$i = \frac{1 - V_A}{4R} = \frac{1}{4R} \left[ 1 - \left( \frac{1}{9} + \frac{4V_O}{9} \right) \right]$$

.. Voltage at non-inverting terminal



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$$V_{+} = 1 - \frac{1}{4R} \times 3R \left[ 1 - \left( \frac{1}{9} + \frac{4V_{o}}{9} \right) \right]$$

$$V_{+} = 1 - \frac{3}{4} \left[ \frac{8}{9} - \frac{4V_{o}}{9} \right]$$

$$V_{+} = 1 - \frac{2}{3} + \frac{V_{o}}{3}$$

$$V_{+} = \frac{1}{3} + \frac{V_{o}}{3}$$

Using virtual short concept

$$V_{+} = V_{-} = \frac{1}{3} + \frac{V_{o}}{3}$$

Applying KCL at inverting node

$$\frac{0 - V_{-}}{3R} = \frac{V_{-} - V_{o}}{3R}$$
$$-V_{-} = V_{-} - V_{o}$$
$$V_{o} = 2V_{-}$$

$$V_o = 2\left[\frac{1}{3} + \frac{V_o}{3}\right]$$

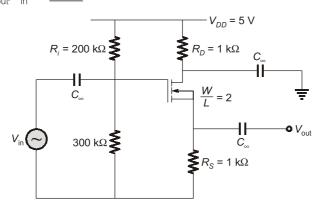
$$3V_o = 2 + 2V_o$$
$$V_o = 2 \text{ Volt}$$



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Q.6 Figure below shows a feedback amplifier constructed using an nMOS transistor. Assume that  $\mu_n C_{ox} = 1$  mA/V², threshold voltage  $V_T = 1$  V and W/L = 2. The bias voltage at the drain terminal is 4 V. The capacitor  $C_{\infty}$  offer zero impedance once at signal frequency. The ratio  $V_{out}/V_{in}$  is \_\_\_\_\_.



Ans. (0.66)

Given: 
$$V_D = 4 \text{ V}, \ \frac{W}{L} = 2, \ V_T = 1 \text{ V}$$

$$\mu_n C_{ox} = 1 \text{ mA/V}^2$$

Given amplifier is common drain amplifier.

Now, 
$$g_{m} = \mu_{n}C_{ox}\frac{W}{L}(V_{GS} - V_{T})$$

$$V_{GS} = V_{G} - V_{S}$$

$$V_{GS} = \frac{5 \times 300}{500} - I_{D}R_{S}$$

$$V_{GS} = 3 - I_{D}R_{S}$$
Also, 
$$I_{D} = \frac{5 - V_{D}}{R_{D}} = \frac{5 - 4}{1} = 1 \text{ mA}$$

$$\therefore V_{GS} = 3 - 1$$

$$V_{GS} = 2 \text{ Volt}$$
Hence, 
$$g_{m} = 1 \times 10^{-3} \times 2[2 - 1]$$

$$g_{m} = 2 \text{ mA/V}$$

.. For common drain amplifier,

$$\frac{V_{\text{out}}}{V_{\text{in}}} = \frac{g_m R_s}{1 + g_m R_s}$$

$$= \frac{2 \times 1}{1 + 2 \times 1}$$

$$\frac{V_{\text{out}}}{V_{\text{in}}} = \frac{2}{3} \text{ V/V}$$
(Taking  $r_{ds} = \infty$ )





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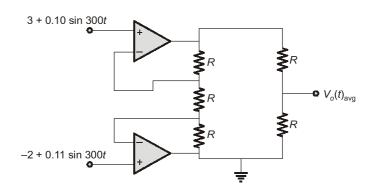


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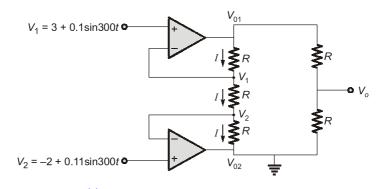
Forenoon Session

# Q.7 For the circuit shown below:



The value of  $V_o(t)_{avg} =$ 

#### Ans. (0.5)



$$V_{O_1} = IR + V_1$$
  
=  $\frac{(V_1 - V_2)}{R}R + V_1 = 2V_1 - V_2$  ...(i)

$$V_{o_{2}} = V_{2} - IR = V_{2} - \frac{(V_{1} - V_{2})}{R} \times R$$

$$= 2V_{2} - V_{1} \qquad ....(ii)$$

$$V_{o} = \frac{V_{o_{1}} \times R}{R + R} + \frac{V_{o_{2}} \times R}{R + R} = \frac{V_{o_{1}} + V_{o_{2}}}{2}$$

$$= \frac{2V_{1} - V_{2} + 2V_{2} - V_{1}}{2} = \frac{V_{1} + V_{2}}{2}$$

$$\stackrel{\checkmark}{\bullet}_{o_{1}}$$

$$V_o = \frac{V_1 + V_2}{2} = \frac{3 + 0.10\sin 300t - 2 + 0.11\sin 300t}{2}$$
$$= \frac{1 + 0.21\sin 300t}{2} = \frac{1}{2} + \frac{0.21}{2}\sin 300t$$
$$= 0.5 + 0.105\sin 300t$$

 $V_{o(avg)} = 0.5 \text{ V}$ 

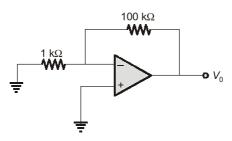
 $V_{o_1}$  R  $V_{o_2}$ 



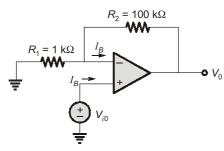
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Q.8 The opamp in the circuit shown is ideal, except that it has an input current of 1 nA and an input offset voltage of 10  $\mu$ V. The resulting worst - case output voltage will be  $\pm$   $\mu$ V.



Ans. (1110)



KCL at inverting terminal:

$$\frac{0 - V_{i0}}{R_1} + \frac{V_0 - V_{i0}}{R_2} = I_B$$

$$\frac{V_0}{R_2} = I_B + \left(\frac{1}{R_1} + \frac{1}{R_2}\right) V_{i0}$$

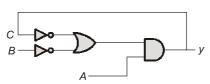
$$V_0 = I_B R_2 + \left(1 + \frac{R_2}{R_1}\right) V_{i0}$$

$$V_0 = 10^{-9} \times 100 \times 10^3 + \left(1 + \frac{100}{1}\right) \times 10 \times 10^{-6}$$

$$V_0 = 1110 \,\mu\text{V}$$



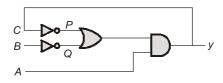
Q.9 Given A = B = 1, initially C = 1, when circuit is turned on. If delays of AND, OR, NOT gates are 10 ns, 10 ns, 5 ns respectively then frequency of steady state oscillations of output y is \_\_\_\_ MHz.



Ans. (20)

Given

$$A = B = 1$$
  
 $C = 1$  (Initially)



From circuit

$$P = \overline{C} = \overline{V}$$

$$Q = \overline{B}$$

Next state of the output

$$y^+ = A \cdot (\overline{B} + \overline{C})$$

A = 1 and B = 1

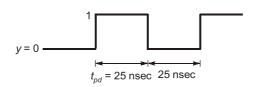
$$y^+ = 1 \cdot (0 + \overline{C}) = 0 + \overline{y} = \overline{y}$$

$$V^+ = \overline{V}$$

Overall propagation delay of circuit,

$$t_{Pd} = 5 + 10 + 10 = 25 \text{ nsec}$$

So, after every 25 nsec the output will toggle.



Frequency of steady state oscillations,

$$f = \frac{1}{50 \times 10^{-9}} = 20 \times 10^6 \text{ Hz}$$
  
 $f = 20 \text{ MHz}$ 



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Q.10 Simplify  $F(w, x, y, z) = \Sigma m(4, 5, 10, 11, 12, 13, 14, 15)$ 

(a)  $x\overline{y} + wy$ 

(b)  $Wx + \overline{W}x\overline{y} + W\overline{x}y$ 

(c)  $Wx + Wy + \overline{x}y$ 

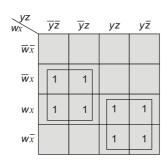
(d)  $\overline{x}y + \overline{w}\overline{y}$ 

Ans. (a)

Given:

 $F(w, x, y, z) = \Sigma m(4, 5, 10, 11, 12, 13, 14, 15)$ 

K-Map:



$$F(w, x, y, z) = x\overline{y} + wy$$

**End of Solution** 

Q.11 Match the following List-1 to List-2.

(1)  $x \oplus x$ 

(P) 1

(2)  $x \oplus \overline{x}$ 

(Q) 0

(3)  $x \oplus 0$ 

(R)  $\bar{x}$ 

(4)  $x \oplus 1$ 

(S) x

Ans. (##)

Truth table for EXOR gate

Inputs		Outpu		
Α	В	Υ		
0	0	0	_	→ Same inputs
1	1	1	*	Different inputs
1	0	1	-	Dillerent inputs
1	1	0	_	→ Same inputs

lr	puts	Output
Α	В	Υ
0	0 —	<del></del> 0
0	1 —	<del></del>
1	0	1
1 1		0

Output Inputs Α В Υ 0 0 0 0 1 1 1 0 -**-** 1 1 1 -**-** 0

If one input of EXOR gate is 0, then the output is same as second input.

If one output of EXOR gate is 1, then the output is the complement of second input.

- $(1) \to (Q)$
- $(2) \rightarrow (P)$
- $(3) \rightarrow (S)$
- $(4) \to (R)$



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Forenoon Session

Q.12 The following excitation table is corresponding to which of the following Flip-flop?

j	Q(t)	Q(t+1)	Input
ĺ	0	0	0
ĺ	0	1	1
	1	0	1
	1	1	0

(a) D-Flip-flop

(b) JK-Flip-flop

(c) SR-Flip-flop

(d) T-Flip-flop

Ans. (d)

When,

Input = 0

Q(t+1) = Q(t)

When,

Input = 1

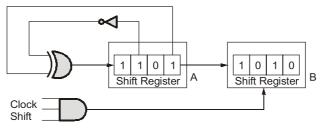
$$Q(t + 1) = \overline{Q}(t)$$

Here, the output toggle for input = 1

So, it is a T-flip-flop.

**End of Solution** 

Q.13 In the circuit shown, the initial binary content of shift register A is 1101 and that of shift register B is 1010. The shift register are +ve edge triggered, and gates have no delay. When the shift control is high, what will be the binary content of the shift register A and B after 4 clock pulse.



- (a) A = 1101, B = 1101
- (b) A = 0101, B = 1101
- (c) A = 1010, B = 1111
- (d) A = 1110, B = 1001

Ans. (b)

	Register A				R	egis	ster	В	
	1	1	0	1		1	0	1	0
Clock 1:	1	1	1	0		1	1	0	1
Clock 2:	0	1	1	1		0	1	1	0
Clock 3:	1	0	1	1		1	0	1	1
Clock 4:	0	1	0	1		1	1	0	1
After 4th clock pulse,									
					<i>A</i> =	0 1	0	1	

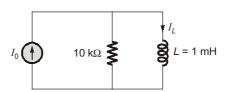
 $B = 1 \ 1 \ 0 \ 1$ 



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The circuit is excited by step current  $I_0u(t)$  at  $t=0^-$ ,  $I_L$  flows  $I_L=\frac{I_0}{5}$ . The minimum Q.14 



Ans. (0.438)

The current equation for inductor

$$\begin{split} i_{L}(t) &= i_{L}(\infty) + \left[i_{L}(0^{+}) - i_{L}(\infty)\right] e^{-t/\tau} \\ i_{L}(0^{+}) &= \frac{I_{o}}{5} \\ i_{L}(\infty) &= I_{o} \\ i_{L}(t) &= I_{o} + \left[\frac{I_{o}}{5} - I_{o}\right] e^{-t/\tau} = I_{o} - \frac{4I_{o}}{5} e^{-t/\tau} \end{split}$$

The inductor current reaches to 99% of its final value.

i.e. 
$$i_L(t) = 0.99 \, I_o$$

$$0.99 \, I_o = I_o \bigg[ 1 - \frac{4}{5} e^{-t/\tau} \bigg]$$

$$\frac{4}{5} e^{-t/\tau} = 0.01$$

$$t = -\tau \ln \bigg( \frac{0.05}{4} \bigg) = 4.382 \tau$$

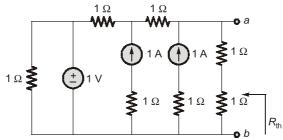
$$\tau = \frac{L}{R} = \frac{1 \times 10^{-3}}{10 \times 10^3} = 10^{-7} \, \mathrm{sec}$$

$$t = 4.382 \times 10^{-7} \, \mathrm{sec}$$

$$t = 0.4382 \, \mu \mathrm{s}$$

**End of Solution** 

Q.15 For the circuit shown below:



The equivalent resistance  $R_{\rm th}$  is \_



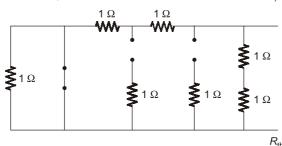
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Ans. (1)

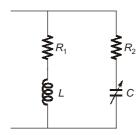
For Thevenin resistance, we need to deactivate all the independent source.



$$R_{\text{th}} = (1+1) \| (1+1) = 2 \| 2$$
  
=  $\frac{2 \times 2}{2+2} = \frac{4}{4} = 1 \Omega$   
 $R_{\text{th}} = 1 \Omega$ 

**End of Solution** 

Q.16 For the circuit shown below:



$$R_1 = R_2 = 2.2 \ \Omega$$

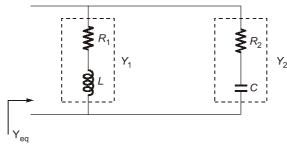
L = 7 mH

and  $\omega = 100 \pi$  rad/sec then the value of C (in mF) is \_\_\_\_\_.

Ans. (1.45)

Given:

$$R_1 = R_2 = 2.2 \Omega$$
  
 $L = 7 \text{ mH}, \omega = 100\pi \text{ rad/sec}$ 



$$Y_{\text{eq}} = Y_1 + Y_2$$
  
=  $\frac{1}{R_1 + j\omega L} + \frac{1}{R_2 - \frac{j}{\omega C}}$ 



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$$= \frac{(R_1 - j\omega L)}{\sqrt{R_1^2 + (\omega L)^2}} + \frac{R_2 + \frac{j}{\omega C}}{\sqrt{R_2^2 + \left(\frac{1}{\omega C}\right)^2}}$$

For resonant frequency,

$$I_{m}\{Y_{eq}\} = 0$$

$$\frac{-j\omega L}{\sqrt{R_{1}^{2} + (\omega L)^{2}}} + \frac{\frac{j}{\omega C}}{\sqrt{R_{2}^{2} + \left(\frac{1}{\omega C}\right)^{2}}} = 0$$

$$\omega L \sqrt{R_{2}^{2} + \left(\frac{1}{\omega C}\right)^{2}} = \frac{1}{\omega C} \sqrt{R_{1}^{2} + (\omega L)^{2}}$$

$$\omega^{2} L^{2} \left(R_{2}^{2} + \frac{1}{\omega^{2} C^{2}}\right) = \frac{1}{\omega^{2} C^{2}} (R_{1}^{2} + \omega^{2} L^{2})$$

$$\omega^{2} L^{2} R_{2}^{2} + \frac{L^{2}}{C^{2}} = \frac{R_{1}^{2}}{\omega^{2} C^{2}} + \frac{L^{2}}{C^{2}}$$

$$\omega^{2} L^{2} R_{2}^{2} = \frac{R_{1}^{2}}{\omega^{2} C^{2}}$$

$$C^{2} = \frac{R_{1}^{2}}{R_{2}^{2}} \cdot \frac{1}{\omega^{4} L^{2}}$$

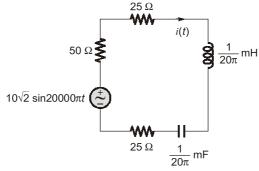
$$C = \frac{R_{1}}{R_{2}} \cdot \frac{1}{\omega^{2} L}$$

$$= \frac{2.2}{2.2} \times \frac{1}{(100\pi)^{2} \times 7 \times 10^{-3}}$$

$$C = 1.45 \times 10^{-3} \text{ F}$$

End of Solution

#### Q.17 For the circuit shown below:



rms value of i(t) is \_\_\_\_\_\_



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Ans. (0.1)

The resonant frequency,

$$\omega_{\rm o} = \frac{1}{\sqrt{LC}} = \frac{1}{\sqrt{\frac{1}{20\pi} \times 10^{-3} \times \frac{1}{20\pi} \times 10^{-3}}}$$

=  $20\pi \times 10^3$  rad/sec

 $\omega_0 = 20000\pi \, \text{rad/sec}$ 

Hence, the circuit is under resonance.

So, the total current,

$$i(t) = \frac{10\sqrt{2}\angle0^{\circ}}{50 + 25 + 25}$$
$$= 0.1\sqrt{2}$$

RMS value of 
$$i(t) = \frac{0.1\sqrt{2}}{\sqrt{2}} = 0.1 \text{ A}$$

$$i(t)|_{rms} = 0.1 \text{ Amp}$$

End of Solution

- Q.18 Force per unit length between two infinitely long parallel conductors, with a gap of 2 cm between them is 10  $\mu$ N/m, when the gap is doubled, then force per unit length will be \_\_\_\_  $\mu$ N/m.
- Ans. (5)

We know that,

Force is inversely proportional to 'd'

$$F \propto \frac{1}{d}$$

$$\frac{F_1}{F_2} = \frac{d_2}{d_1}$$

$$\frac{10}{F_2} = \frac{4 \text{ cm}}{2 \text{ cm}}$$

$$F_2 = \frac{10}{2} = 5 \,\mu\text{N/m}$$



**Forenoon Session** 

Q.19 If 
$$f(x) = x \sin\left(\frac{1}{x}\right)$$
; then  $\lim_{x\to 0} f(x) = ?$ 

Ans. (0)

$$\lim_{x \to 0} x \sin \frac{1}{x} = \lim_{x \to 0} \frac{\sin \frac{1}{x}}{\frac{1}{x}}$$

$$\lim_{V \to \infty} \frac{\sin y}{V} = \frac{\text{finite}}{\infty} = 0$$

**End of Solution** 

Q.20 If 
$$f(z) = i \frac{1-z}{1+z}$$
 then  $f^{-1}(z)$  maps real axis to

- (a) unit circle with center not at origin
- (b) real axis
- (c) imaginary axis
- (d) unit circle with center at origin

Ans. (d)

Given:

$$W = U + iV = \frac{i(1-Z)}{(1+Z)}$$

$$= \frac{i - ix - i^2 y}{1+x+iy}$$

$$= \frac{i(1-x) + y}{(1+x) + iy} \times \frac{(1+x) - iy}{(1+x) - iy}$$

$$u + iv = \frac{y(1-x) + y(1+x)}{(1+x)^2 + y^2} + i\frac{(1-x^2) - y^2}{(1+x)^2 + y^2}$$

To find image of real axis is w-plane

We take

$$V = 0$$

$$\Rightarrow \frac{(1-x^2)-y^2}{(1+x)^2+y^2} = 0$$

$$\Rightarrow 1 - x^2 - y^2 = 0$$

$$\Rightarrow \qquad x^2 + y^2 = 1$$

:. Under the mapping of  $f^{-1}(z)$  real axis in w-plane maps to unit circle in z-plane.



Exam held on:

11-02-2023 **Forenoon Session** 

Q.21 
$$A = \begin{bmatrix} 1 & 4 \\ -3 & \alpha \\ \beta & 6 \end{bmatrix}$$
 If Rank of matrix A is 1 then the ratio of  $\frac{\alpha}{\beta}$  is \_\_\_\_\_.

Ans. (-8)

$$\Rightarrow$$

*:*.

$$\rho(A_{3\times 2}) = 1$$

$$C_2 = KC_1$$

$$4 = K(1)$$

$$C_2 = KC$$

$$\alpha = -3K$$
 and  $6 = K\beta$ 

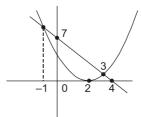
$$\frac{\alpha}{\beta} = -\frac{1}{2}K^2 = -\frac{1}{2} \times 4^2 = -8$$

End of Solution

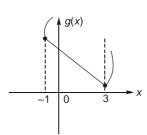
**Q.22** 
$$g(x) = \max\{(x-2)^2, -2x + 7\}; x \in (-\infty, \infty).$$
 Minimum value of  $g(x) =$ \_\_\_\_\_.

Ans. (1)

$$g(x) = \max \{(x-2)^2, -2x + 7\}$$



Then g(x)



From the figure:

Minimum value of g(x) is at

$$x = 3$$

$$g_{min}(x) = g(3)$$
  
=  $(3 - 2)^2$  or  $(-2x + 7)_{x=3}$   
= 1



**Forenoon Session** 

Q.23 If X is a discrete random variable

	Χ	0	1	2
İ	Р	0.25	0.5	?

then value of  $[E(X) - E(X^2)] = \underline{\hspace{1cm}}$ 

Ans. (-0.5)

x is discrete R.V

W.K.T: 
$$\sum_{i=1}^{n} P(x = x_i) = 1$$

$$\Rightarrow 0.25 + 0.5 + k = 1$$

$$k = 1 - 0.75$$

$$k = 0.25$$

$$\therefore E(x) = \sum_{i} x_i P_i$$

$$= 0 \times 0.25 + 1 \times 0.5 + 2 \times 0.25$$

$$0.5 + 0.5$$

$$E(x) = 1$$

$$E(x^2) = \sum_{i} x_i^2 P_i$$

$$= 0^2 \times 0.25 + 1^2 \times 0.5 + 2^2 \times 0.25$$

$$= 0 + 0.5 + 1$$

$$E(x^2) = 1.5$$

 $E(x) - E(x^2) = 1 - 1.5 = -0.5$ 

End of Solution

Choose solution 'S' for set of equations Q.24

$$x - 2y + z = 0$$
$$x - z = 0$$

(a) 
$$S = \left\{ \alpha \begin{pmatrix} 1 \\ 1 \\ 1 \end{pmatrix} + \beta \begin{pmatrix} 1 \\ 0 \\ 1 \end{pmatrix} \middle| \alpha, \beta \in R \right\}$$
 (b)  $S = \left\{ \alpha \begin{pmatrix} 1 \\ 1 \\ 1 \end{pmatrix} \middle| \alpha \in R \right\}$ 

(b) 
$$S = \begin{cases} \alpha \begin{pmatrix} 1 \\ 1 \\ 1 \end{pmatrix} | \alpha \in R \end{cases}$$

(c) 
$$S = \left\{ \alpha \begin{pmatrix} 1 \\ 1 \\ 1 \end{pmatrix} + \beta \begin{pmatrix} 2 \\ 1 \\ 1 \end{pmatrix} \middle| \alpha, \beta \in R \right\}$$
 (d)  $S = \left\{ \alpha \begin{pmatrix} 1 \\ 0 \\ 1 \end{pmatrix} \middle| \alpha \in R \right\}$ 

(d) 
$$S = \begin{cases} \alpha \begin{pmatrix} 1 \\ 0 \\ 1 \end{pmatrix} \middle| \alpha \in R \end{cases}$$



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Ans. (b)

Given: 
$$x - 2y + z = 0$$
  
 $x - z = 0$ 

Solution is given by solving also equations.

$$\frac{x}{2} = \frac{y}{2} = \frac{z}{2} = K$$

$$\Rightarrow \qquad x = K, \ y = K, \ z = K$$

$$\therefore \begin{bmatrix} x \\ y \\ z \end{bmatrix} = K \begin{bmatrix} 1 \\ 1 \\ 1 \end{bmatrix}, K \in R$$

End of Solution

- Q.25 In a P-i-n photodiode, a pulse of light containing  $8\times 10^{12}$  incident photons at wavelength  $\lambda_0=1.55~\mu m$  gives rise to an average  $4\times 10^{12}$  electrons collected at the terminals of the device. The quantum efficiency of the photodiode at its wavelength is
  - (a) 54.2

(b) 62.5

(c) 80

(d) 50

Ans. (d)

quantum efficiency, 
$$\eta = \frac{\text{Electrons out}}{\text{Photons incident}} = \frac{4 \times 10^{12}}{8 \times 10^{12}} = 0.5$$

$$\therefore \qquad \qquad \eta = 50\%$$

End of Solution

- Q.26 LED is operated under \_\_\_\_\_ and photodiode is operated under \_\_\_\_\_.
  - (a) Forward biased and Reverse biased
  - (b) Forward biased and Forward biased
  - (c) Reverse biased and Reverse biased
  - (d) Reverse biased and Forward biased

Ans. (a)

**End of Solution** 

Q.27 A silica glass fibre has a core refractive index of 1.47 and a cladding refractive index of 1.44. If the cladding is completely stripped out and core is dipped in water having a refractive index of 1.33, the numerical aperture of modified fiber is \_\_\_\_\_.



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Ans. (0.62)

Numerical Aperture, N.A = 
$$\sqrt{n_{\text{core}}^2 - n_{\text{water}}^2} = \sqrt{1.47^2 - 1.33^2}$$
  
N.A = 0.62

End of Solution

Q.28 Number of times Nyquist plot of 
$$G(s)H(s) = \frac{(s-1)(s-2)}{2(s+1)(s+2)}$$
 encircles the origin is \_\_\_\_\_.

Ans. (2)

$$G(s)H(s) = \frac{(s-1)(s-2)}{2(s+1)(s+2)}$$

 $\therefore$  Number of poles which are in RHS, P=0Number of zeros which are in RHS, Z=2(Here it is asking for number of encirclement around the origin).

∴ Encirclement, 
$$N = P - Z$$
  
= 0 - 2  
 $N = -2$ 

.. 2 clockwise encirclement.

End of Solution

**Q.29** Number of zeros of  $s^3 + 2s^2 + 5s + 80$  in right half plane is \_\_\_\_\_.

Ans. (2)  $s^3 + 2s^2 + 5s + 80$ 

 $s^3 + 2s^2 + 5s + 8$ Forming RH array

$$s^{3}$$
 1 5  
 $s^{2}$  2 80  
 $s^{1}$   $\frac{10-80}{2} = -35$   
 $s^{0}$  80

∴ There are two sign change. Hence, 2 open loop zeros in RHS.

**End of Solution** 

**Q.30** If 
$$G(s) = \frac{2(1-s)}{(1+s)^2}$$
 then phase margin of  $G(s) =$ \_\_\_\_\_ degree.

Ans. (0)

$$G(s) = \frac{2(1-s)}{(1+s)^2}$$



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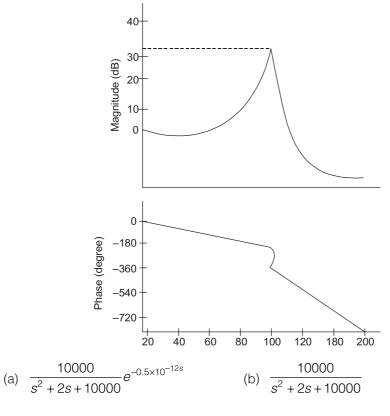
11-02-2023
Forenoon Session

Phase margin is calculated at gain crossover frequency  $\omega_{gc}$ .  $\omega_{gc}$  is the frequency at which the magnitude of  $G(j\omega)H(j\omega)$  is 1.

$$\begin{aligned} \left| G(j\omega) H(j\omega) \right|_{\omega = \omega_{gc}} &= \frac{2\sqrt{1 + \omega_{gc}^2}}{\left(1 + \omega_{gc}^2\right)} = \frac{2}{\sqrt{1 + \omega_{gc}^2}} = 1 \\ & 2 = \sqrt{1 + \omega_{gc}^2} \\ & 4 = 1 + \omega_{gc}^2 \\ & \omega_{gc} = \sqrt{3} \text{ rad/sec} \\ & \text{Now} \qquad \phi \text{ at } \omega_{gc} = \tan^{-1}[-\omega_{gc}] - 2 \tan^{-1}[\omega_{gc}] \\ & = -3 \tan^{-1}[\omega_{gc}] \\ & = -3 \tan^{-1}(\sqrt{3}) \\ & = -180^{\circ} \\ & \therefore \qquad \text{Phase margin} = 180 + \phi \text{ at } \omega_{gc} \\ & = 180^{\circ} - 180^{\circ} = 0^{\circ} \end{aligned}$$

**End of Solution** 

Q.31 Magnitude and phase plot shown in figure. Match with the transfer function.



(c) 
$$\frac{10000}{s^2 + 2s + 10000}e^{-0.05s}$$
 (d)  $\frac{100}{s^2 + 2s + 1000}$ 



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11-02-2023 **Forenoon Session** 

#### Ans. (c)

From the figure it is clear that there should be phase lag.

Hence, option (a) and (c) could be answer.

From the figure, at  $\omega = 100$  rad/sec, Phase is  $-360^{\circ}$ .

Now considering option (c) only delay element

$$e^{-0.05s} = e^{-0.05 \times 100j} = e^{-j5}$$

Hence, phase is = -5 radians =  $-286.5^{\circ}$ 

and from the rest of transfer function  $\frac{1000}{s^2 + 2s + 10000}$ 

at  $\omega = \omega_n = 100 \text{ rad/sec phase is } -90^\circ$ .

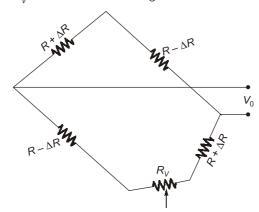
$$= -286.5 - 90^{\circ}$$

$$= -376.5^{\circ}$$

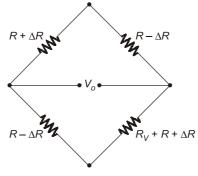
Hence, option (c) satisfy the graph.

End of Solution

Q.32 Initially bridge was balanced and after applying force resistances are changed then find minimum value of  $R_V$  to balance the bridge.



Ans. (##)



$$V_o = E \left[ \frac{R + \Delta R}{R + \Delta R + R - \Delta R} - \frac{R - \Delta R}{R - \Delta R + (R_V + R + \Delta R)} \right]$$

$$V_o = E \left[ \frac{R + \Delta R}{2R} - \frac{R - \Delta R}{2R + R_V} \right]$$



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**Forenoon Session** 

For a balanced bridge 
$$\Rightarrow V_b = 0$$
 
$$0 = E \bigg[ \frac{R + \Delta R}{2R} - \frac{R - \Delta R}{2R + R_V} \bigg]$$
 
$$\frac{R + \Delta R}{2R} = \frac{R - \Delta R}{2R + R_V} \Rightarrow 2R + R_V = 2R \bigg[ \frac{R - \Delta R}{R + \Delta R} \bigg]$$
 
$$R_V = 2R \bigg[ \frac{R - \Delta R}{R + \Delta R} - 1 \bigg] = 2R \bigg[ \frac{R - \Delta R - R - \Delta R}{R + \Delta R} \bigg]$$
 
$$= 2R \bigg[ -\frac{2\Delta R}{R + \Delta R} \bigg] = \frac{-4R\Delta R}{R \bigg[ 1 + \frac{\Delta R}{R} \bigg]} = -\frac{4\Delta R}{1 + \frac{\Delta R}{R}} \qquad \bigg( \frac{\Delta R}{R} \text{ neglected} \bigg)$$
 
$$= -4\Delta R \Rightarrow \text{ minimum value}$$

**End of Solution** 

The Laplace transform of the continuous time signal  $x(t) = e^{-3t}u(t-5)$  is \_\_\_\_\_\_. u(t)Q.33 denotes the continuous time unit step signal.

(a) 
$$\frac{e^{-5(s-3)}}{s-3}$$
, Real $\{s\} > -3$ 

(b) 
$$\frac{e^{-5(s+3)}}{s+3}$$
, Real $\{s\} > -3$ 

(c) 
$$\frac{e^{-5(s-3)}}{s+3}$$
, Real $\{s\} > -3$  (d)  $\frac{e^{-5s}}{s+3}$ , Real $\{s\} > -3$ 

(d) 
$$\frac{e^{-5s}}{s+3}$$
, Real $\{s\} > -3$ 

Ans. (b)

We have,

$$x(t) = e^{-3t}u(t-5)$$

We know,

$$u(t) \rightarrow \frac{1}{s}$$

$$u(t-5) \to \frac{1}{s} \cdot e^{-5s}$$

$$e^{-3t}u(t-5) \to \frac{1}{(s+3)} \cdot e^{-5(s+3)}$$

$$X(s) = \frac{e^{-5(s+3)}}{s+3};$$
 Re(s) > -3



If y(t) = x(4t) where x(t) is a continuous time periodic signal with fundamental period Q.34 of 100 sec, then the fundamental period of y(t) is \_\_\_\_\_ sec.

Ans. (25)

$$x(t) = \sin(\omega_o t)$$

$$T = 100 \sec 0.00$$

$$\omega_o = \frac{2\pi}{T} = \frac{2\pi}{100}$$
 rad/sec

$$y(t) = x(4t) = \sin(4\omega_0 t)$$

$$\omega' = 4\omega_o = 4 \times \frac{2\pi}{100}$$

$$\frac{2\pi}{T'} = 4 \times \frac{2\pi}{100}$$

$$T' = 25 \text{ sec}$$

End of Solution

For LTI system with input x(t) and output y(t) have transfer function  $H(s) = \frac{Y(s)}{X(s)} = \frac{s - \pi}{s + \pi}$ Q.35 if  $y(t) = \sin \pi t$  then x(t) is

(a) 
$$\cos \left[ \pi t + \frac{\pi}{4} \right]$$

(b) 
$$\sin \left[ \pi t - \frac{\pi}{2} \right]$$

(c) 
$$\sin\left[\pi t + \frac{\pi}{2}\right]$$

(d) 
$$\sin \pi t$$

Ans. (b)

We know,

$$x(t) = \frac{H(s)}{A \sin(\pi t + \theta)}$$
 LTI system 
$$y(t) = A' \sin(\pi t + \phi)$$

where,

$$A' = A \cdot \big| H(\omega) \big|_{\omega = \omega_{\mathcal{O}}}$$

$$\phi = \left. \theta + \angle H(\omega) \right|_{\omega = \omega_0}$$

We have,

$$\omega_{\scriptscriptstyle O}=\pi,~A'=1,~\phi=0$$

Now,

$$|H(\omega)|_{\omega=\pi} = \frac{\sqrt{\omega^2 + \pi^2}}{\sqrt{\omega^2 + \pi^2}} = 1$$

$$\angle H(\omega)|_{\omega=\pi} = 180^{\circ} - \tan^{-1}\left(\frac{\omega}{\pi}\right) - \tan^{-1}\left(\frac{\omega}{\pi}\right)$$
$$= 180^{\circ} - 45^{\circ} - 45^{\circ}$$

= 90°

$$A = 1$$
$$\theta = -90^{\circ}$$

Now,

$$x(t) = \sin(\pi t - 90^\circ) = \sin\left(\pi t - \frac{\pi}{2}\right)$$



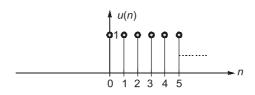
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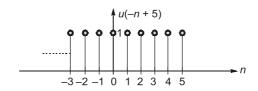
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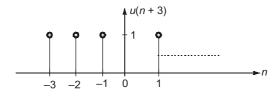
Q.36 Consider the discrete time signal 
$$x(n) = u(-n + 5) - u(n + 3)$$
 where  $u(n) = \begin{cases} 1; & n \ge 0 \\ 0; & n < 0 \end{cases}$ .

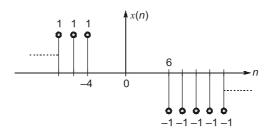
The smallest 'n' for which  $x(n) = 0$  is \_\_\_\_\_\_.

Ans. (-3)









Smallest 'n' for which x[n] = 0 $\therefore$  n = -3



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- 58(t-4) where 8(t) is the
- Q.37 The impulse response of an LTI system is  $h(t) = \delta(t) + 0.5\delta(t-4)$ , where  $\delta(t)$  is the continuous time unit impulse signal. If the input signal  $x(t) = \cos\left(\frac{7\pi}{4}t\right)$  then output is

(a) 
$$0.5\sin\left(\frac{7\pi}{4}t\right)$$

(b) 
$$1.5\sin\left(\frac{7\pi}{4}t\right)$$

(c) 
$$0.5\cos\left(\frac{7\pi}{4}t\right)$$

(d) 
$$1.5\cos\left(\frac{7\pi}{4}t\right)$$

Ans. (c)

We have;

$$h(t) = \delta(t) + 0.5\delta(t-4)$$

Taking FT;

$$H(\omega) = 1 + 0.5e^{-j4\omega}$$

at

$$\omega = \frac{7\pi}{4}$$

$$H(\omega) = 1 + 0.5e^{-j4\left(\frac{7\pi}{4}\right)}$$
  
= 1 + 0.5 $e^{-j7\pi}$   
= 0.5

$$\therefore \qquad |H(\omega)|_{\omega = \frac{7\pi}{4}} = 0.5$$

$$\angle H(\omega)|_{\omega=\frac{7\pi}{4}} = 0$$

$$x(t) = A \sin(\omega_0 t + \theta)$$
LTI system
$$H(\omega)$$

$$y(t) = A' \sin(\omega_0 t + \phi)$$

$$A' = A_0 |H(\omega)|_{\omega = \omega_0}$$

$$\phi = \left. \theta + \angle H(\omega) \right|_{\omega = \omega_0}$$

$$y(t) = 0.5\cos\left(\frac{7\pi}{4}t\right)$$

End of Solution

- **Q.38** A continuous real valued signal x(t) has finite positive energy and x(t) = 0, t < 0. From the list given below select all the signal whose continuous time Fourier transform is purely imaginary.
  - (a) x(t) + x(-t)

(b) j(x(t) + x(-t))

(c) x(t) - x(-t)

(d) j(x(t) - x(-t))



#### Ans. (b, c)

If x(t) is real, then its Fourier transform will be complex conjugate.

i.e., 
$$x(t) \longrightarrow X(\omega)$$

$$x(-t) \longrightarrow X(-\omega) = X^{*}(\omega)$$

Option (a) 
$$\rightarrow x(t) + x(-t) \xleftarrow{FT} X(\omega) + X(-\omega) = X(\omega) + X^*(\omega) \Rightarrow \text{Real}$$

Option (b) 
$$\rightarrow$$
  $f[(x(t) + x(-t))]$ 

$$= \int [X(\omega) + X(-\omega)]$$

... Taking Fourier Transform

$$= j[X(\omega) + X(-\omega)]$$
$$= i(X(\omega) + X^*(\omega))$$

= Purely Imaginary

Option (c) 
$$\rightarrow x(t) - x(-t) \xleftarrow{F.T} X(\omega) - X(-\omega) = X(\omega) - X^*(\omega) \Rightarrow \text{Purely imaginary}$$

Option (d) 
$$\rightarrow j[x(t) - x(-t)] \xleftarrow{F.T} j[X(\omega) - X^*(\omega)] \Rightarrow \text{Purely Real}$$

End of Solution

Q.39 If 
$$F(z) = \frac{1}{1-z}$$
 expanded as power series around  $z = 2$  would result in  $f(z) = \sum_{k=0}^{\infty} a_k (z-2)^k$  with region of convergence  $|z-2| < 1$  then value of  $a_k$  is

(a) 
$$\left(\frac{1}{2}\right)^k$$

(b) 
$$(-1)^k$$

(c) 
$$(-1)^{k+1}$$

(d) 
$$\left(\frac{1}{z}\right)^{k+1}$$

#### Ans. (c)

Power series expansion for

$$\frac{1}{1+x} = 1 - x + x^2 - x^3 + x^4 - \dots$$
 when  $|x| < 1$ 

*:*.

Now, given that, 
$$F(z) = \frac{1}{1-z} = -\left[\frac{1}{1+(z-2)}\right]$$
$$= -[1-(z-2)+(z-2)^2-(z-2)^3+\dots],$$

when 
$$|z - 2| < 1$$

$$= -\sum_{K=0}^{\infty} (-1)^{K} (z-2)^{K}, \quad |z-2| < 1$$

$$= \sum_{K=0}^{\infty} (-1)^{K+1} (z-2)^K, \quad |z-2| < 1$$

$$= \sum_{K=0}^{\infty} a_K (z-2)^K, \quad |z-2| < 1$$

$$a_K = (-1)^{K+1}$$